COMBINING SYNERGETIC CONTROL AND SUPER TWISTING ALGORITHM TO REDUCE THE ACTIVE POWER UNDULATIONS OF DOUBLY FED INDUCTION GENERATOR FOR DUAL-ROTOR WIND TURBINE SYSTEM

Aim. This work presents the amelioration of direct power control using synergetic-super twisting algorithms for asynchronous generators integrated into dual-rotor wind turbine systems. Method. The main role of the direct power control is to control the active and reactive powers and reduce the harmonic distortion of stator current of asynchronous generator for variable speed dual-rotor wind turbine systems. The traditional strategy is more attractive due to its high efficiency and simple algorithm. Super twisting algorithms are a non-linear command strategy, characterized by robustness against the parameters change or disturbances, it gives a good power quality under different conditions such as changing generator parameters. Novelty. Synergetic-super twisting algorithms are designed. Synergetic-super twisting algorithms construction is based on synergetic command and super twisting algorithms in order to obtain a robust control strategy and a fast system with acceptable precision. We use in our study a 1.5 MW asynchronous generator integrated to dual-rotor wind turbine system in order to regulate the active and reactive powers. Results. As shown in the results figures using synergetic-super twisting algorithms the ameliorate performances especially minimizes the torque, active and reactive power undulations, and reduces harmonic distortion of stator current (THD = 0.19 %) compared to traditional strategy. References 40, tables 2, figures 28.

Key words: super twisting algorithm, synergetic command, asynchronous generator, direct active and reactive power command.

Introduction. Nowadays, the increasing demand for electrical energy, its sources and the ever-increasing risks of increasing global warming. On the other hand, the financial cost and the difficulty of implementation and control increase the possibility of not using wind energy, which could cause an increase in the demand for the use of non-renewable resources such as gas, for example, and thus an increase in the emission of CO₂. Also, an increase in the cost of producing electrical energy, which leads to imposing taxes on electric energy consumption.

In the field of electric power generation, there are several electric generators used in generating electric energy using wind energy, for example, asynchronous and synchronous generators.

In our work, we will study the asynchronous generator (ASG) in generating electrical energy using a renewable source. Among the advantages of using an ASG is that it is solid and simple to command, unlike other generators [1]. In the industrial field, there are several types of command methods for electrical machines, especially electrical generators. Among these methods, we mention direct torque command [2], direct power command [3], hybrid command [4], and artificial intelligence use command [5]. Direct power command is...
among the other methods used. This method has its pros and cons just like the other methods, an easy and uncomplicated way, all generators can be converted. Among its disadvantages, we find ripples in the active and reactive powers, which are the most prominent negatives that characterized them. There are several scientific studies in this field that have concluded that fluctuations in reactive and active power oscillations can be reduced by using modern technologies such as fuzzy logic [6], neural networks [7], neuro-fuzzy command [8], sliding mode control (SMC) technique [9], genetic algorithm [10], synergetic control (SYC) [11], super twisting algorithm (STA) [12], etc.

The sliding mode is a particular operating mode of variable structure systems. It is considered one of the simplest approaches for controlling nonlinear systems and systems with an imprecise model. This command has the following characteristics [13]:

- the response of the system is insensitive and robust to variations in certain parameters and the effects of load disturbances and disturbances;
- it suffices to know a terminal for \( u(t) \) which simplifies the adjustment;
- the choice of the switching surface is fairly free;
- the order is softened by the presence of the equivalent order, which can be deleted at the cost of an increase of \( u(t) \).

The sliding surface \( S \) is a scalar function such that the variable to be adjusted slides on this surface. The purpose of the command is to keep the surface at zero. The main drawback of the command in higher-order sliding mode lies in the need to know the state variables and their derivatives. A sliding regime of order \( r \) (noted \( r \)-sliding) acts on the surface and its \((r-1)\) first successive derivatives to time. The objective is to force the system to evolve not only on the surface but also on its \((r-1)\) first successive derivatives and to keep the sliding set at zero:

\[
S = \dot{S} = \ddot{S} = \cdots = (S^{(r-1)}) = 0 ,
\]

with \( r \) designates the relative degree of the system, and its \((r-1)\) first successive derivatives with respect to time.

STA algorithm is a kind of high-order SMC technique. It is characterized by simplicity and durability compared to some techniques. This method was proposed by the Levant in 1993 [14]. This method has been applied in several fields [15-19]. Furthermore, the SYC method is also applied. It tries to overcome the problem of controlling the power converter by using the internal dynamic characteristics of the system, the most important advantages of this approach are order reduction, decoupling design procedure, and insensitivity to parameter changes [20]. On the other hand, this method reduces the vibrations present in the sliding command and improves the stability of the system [21].

A new nonlinear control has been proposed in this paper. This proposed nonlinear control is based on STA algorithms and synergetic control theory.

The aim of this work is the improve the performance of direct reactive and active power control (DRAPC) using synergetic-super twisting algorithms (SYSTA) for ASG-based dual-rotor wind power (DRWP) system under variable speed wind and also to reduce fluctuations in torque, current and active power.

This method is called SYSTA, and it is the product of a marriage of properties of both synergetic control and STA algorithms. This method can be applied to all controls without exception, and it has provided very satisfactory results compared to the classical method.

Model of DRWP. Traditionally, the applied systems of wind turbine systems can be classified into variable speed (VS) and fixed speed turbines (FST). The VS turbine systems (VSTSs) are now more often applied than the systems with FST. The main advantages of VSTSs are: increasing the production of wind power, the ability to achieve maximum power conversion efficiency, and reduction of mechanical stresses. On the other hand, the DRWP is a wind turbine used to generates electrical power. The DRWP system has been proposed as new wind energy, as shown in Fig. 1.

![Fig. 1. Block diagram of DRWP with a ASG](image)

The aerodynamic torque of the auxiliary rotor is [22]:

\[
T_A = \frac{1}{2 \cdot \lambda_A^3} \cdot \rho \cdot \pi \cdot R_A^5 \cdot C_p \cdot w_A^2 ,
\]

and aerodynamic torque of the main rotor is:

\[
T_M = \frac{1}{2 \cdot \lambda_M^3} \cdot \rho \cdot \pi \cdot R_M^5 \cdot C_p \cdot w_M^2 ,
\]

where \( \lambda_A \), \( \lambda_M \) are the tip speed ration of the auxiliary and main rotors; \( R_m \), \( R_A \) are the blade radius of the main and auxiliary rotors, \( \rho \) is the air density; \( w_A \), \( w_M \) are mechanical speed of the auxiliary and main rotors; \( C_p \) is the power coefficient.

The tip speed ratio of the auxiliary rotor (AR) is:

\[
\lambda_A = \frac{w_A \cdot R_A}{V_1} ,
\]

and the tip speed ratio of the main rotor (MR) is:

\[
\lambda_M = \frac{w_M \cdot R_M}{V_M} ,
\]

where \( V_M \) is the speed of the unified wind on main rotor and \( V_1 \) is the wind speed on an auxiliary wind turbine (AWT).

The total aerodynamic torque of DRWP \( (T_T) \) is the sum of AR torque \( (T_A) \) and the MR torque \( (T_M) \):

\[
T_{DRWP} = T_T = T_A + T_M .
\]
The wind speed on the auxiliary and main turbines is the essential element to calculating the tip speed ratio. Equation (6) represents the wind speed in the main turbine [23]

\[ V_s = V'_i \left( 1 - \frac{1 - \sqrt{(1 - C_T)\beta}}{2} \left( 1 + \frac{2 \cdot x}{\sqrt{1 + 4 \cdot x^2}} \right) \right) \]  

where \( x \) is the non-dimensional distance from the auxiliary rotor disk, \( V'_i \) is the velocity of the disturbed wind between rotors at point \( x \); \( C_T \) is the trust coefficient, which is taken 0.9 [24]. The distance between the main and the auxiliary turbines is 15 m.

The \( C_p \) is given as:

\[ C_p(\lambda, \beta) = \frac{1}{\lambda + 0.08 \beta - \beta^2 + 1}, \]  

where \( \lambda \) is the tip speed ratio; \( \beta \) is pitch angle.

**Synergetic-super twisting algorithm.** A system with variable structure is a system whose structure changes during its operation, it is characterized by the choice of a structure and switching logic. This choice allows the system to switch from one structure to another at any time. Moreover, such a system can have new properties which do not exist in every structure.

In the control of systems with variable structure by sliding mode, the state trajectory is brought to a surface, then using the switching law, it is forced to stay in the vicinity of this surface; this latter is called surface sliding movement and the movement along which occurs is called sliding movement [25].

During the last century, many nonlinear methods have been proposed for controlling electrical machines. Among the most famous of them we find control by slip control and this is due to the simplicity of the method and durability. Recently a new theory has appeared called synergetic control [26]. This method is more simple and uncomplicated based on the area derivation calculation. The SYC theory is one of the new methods of robust control [27]. It is characterized by its external disturbance rejection capabilities, simplicity of design, and the global stability assurance of the system [28]. The SYC method is a strategy quite close to the SMC strategy in the sense that it forces the system to evolve with a dynamic pre-chosen by the designer. This novel technique does not require the linearization of the model and explicitly uses a nonlinear model for the synthesis of the control. Also, the SYC method eliminates more the chattering phenomenon compared than the SMC strategy [29].

Equation (13) illustrates the principle of the SYC method, as it depends on the derivation of the surface

\[ TS(x) + S(x) = 0, \]  

where \( T \) is a speed of convergence of surfaces to the intersection of manifolds \( S = 0 \).

The following to ensure the stability of SYC method: \( S(0) = 0, S(x) > 0 \) for all \( x \neq 0 \).

The solution of Eq. (8) is given by:

\[ S(t) = S_0 e^{\frac{t}{T}}. \]  

Basically, STA design follows two steps to implement. In the STA strategy, the command input applies on the second-order derivative of the sliding surface, reverses the SMC it acts on the first derivative of the sliding surface [30]. The command input of the STA method comprises two inputs as (10)

\[ w(t) = w_1 + w_2, \]  

where:

\[ w_1(t) = \lambda_1 \sqrt{S} \cdot \text{sign}(S), \]  

\[ w_2(t) = \lambda_2 \int \text{sign}(S) dt. \]  

The designed strategy has the same objective as the STA and SYC method, it will force the state trajectory to operate on the surface \( S = 0 \). The surface is selected according to system constraints. The proposed method is a combination of the STA and SYC method. This proposed controlled named SYSTA algorithm, where this controller is a simple structure and more robust compared to SYC and STA techniques. Our goal for this controller is to minimize more and more the active and reactive power undulations.

Equation (13) illustrates the principle of the SYSTA controllers

\[ u(t) = u_1(t) + u_2(t). \]  

where the \( u_1(t) \) and \( u_2(t) \) represent the STA method and the synergetic command, respectively

\[ u_1(t) = \alpha_1 \sqrt{S(t)} \cdot \text{sign}(S(t)) + \alpha_2 \int \text{sign}(S(t)) dt, \]  

\[ u_2(t) = \alpha \frac{dS(t)}{dt} + S(t). \]  

The command input of the designed SYSTA method is obtained as (16)

\[ u(t) = \alpha_1 \sqrt{S(t)} \cdot \text{sign}(S(t)) + \alpha_2 \int \text{sign}(S(t)) dt + \]  

\[ + \alpha_3 \frac{dS(t)}{dt} + S(t), \]  

where the tuning constants \( \alpha, \alpha_1 \) and \( \alpha_2 \) are used to tune the SYSTA technique to smoothen the regulator.

This is the design process using the SYSTA controller for the DRAPC method. On the other hand, Fig. 2 shows a block diagram representation of the SYSTA technique for DRAPC command in DRWP systems.

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**Fig. 2. Structure of the command law of the proposed SYSTA technique**

This designed command is used in this work for minimizing electromagnetic torque, rotor current, active power, reactive power, and rotor flux undulations in an ASG-based DRWP system using the DRAPC method which the inverter was controlled by the modified space vector modulation (SVM) technique.

**DRAPC technique.** The principle of the DRAPC strategy is the direct regulation of the reactive and active...
powers of the generator, by applying different voltage vectors to the inverter, which determines its state. The two controlled variables are active and reactive powers which are usually controlled by hysteresis comparators. The idea is to keep the reactive power quantities and the active powers within these hysteresis bands. The output of these regulators determines the optimum voltage vector to be applied at each switching instant [31].

This technique involves the operation of the inverter at two standard levels with a variable control frequency which is sometimes high and incompatible with high power applications due to the level of switching losses.

Two command techniques have been used to implement DRAPC commands:
• command by a lookup table;
• command by a SVM technique.

The purpose of a DRAPC strategy is to keep the reactive power and active power modulus within the hysteresis bands by choosing the output voltage of the inverter. When the active power or reactive power modulus reaches the upper or lower limit of the hysteresis, an appropriate voltage vector is applied to bring the relevant magnitude back from its hysteresis.

Implementing DRAPC commands:
• voltage vectors of the rotor inverter, which rotor flux \( \Psi_r \)
• active power and the reactive power which are controlled by hysteresis regulators. In a DRAPC strategy, it is preferable to work with a high calculation frequency to reduce the reactive and active power oscillations caused by the regulators [32]. A voltage inverter achieves seven distinct positions in the phase plane, corresponding to the eight sequences of the voltage vector at the output of the rotor inverter [33].

The reactive power \( Q_s \) is expressed as a function of the quadrature rotor flux and the direct rotor flux as follows:

\[
Q_s = -\frac{3}{2} \left( \frac{V_s}{\sigma L_s} \right) \cdot \Psi_{r\alpha} - \frac{V_s}{\sigma L_s \cdot L_r} \cdot \Psi_{r\beta}. \tag{17}
\]

The active power command depends directly on the control of the rotation of the rotor flux vector.

\[
P_s = -\frac{3}{2} \frac{L_m}{\sigma L_s \cdot L_r} \left( \frac{V_s}{\sigma L_s} \Psi_{r\beta} \right). \tag{18}
\]

The rotor flux \( \Psi_r \) can be estimated from measurements of the stator current and voltage of the generator.

From the equation:

\[
\Psi_r = \int_{0}^{\Delta t} (V_{r} - R_{r} \cdot i_{r}) \, dt, \tag{19}
\]

we obtain the components \( \alpha \) and \( \beta \) of the vector \( \Psi_r \):

\[
\begin{align*}
\Psi_{r\alpha} &= \int_{0}^{\Delta t} (V_{r\alpha} - R_{r} \cdot i_{r\alpha}) \, dt; \\
\Psi_{r\beta} &= \int_{0}^{\Delta t} (V_{r\beta} - R_{r} \cdot i_{r\beta}) \, dt.
\end{align*} \tag{20}
\]

The rotor flux amplitude is given by:

\[
\Psi_r = \sqrt{\Psi_{r\alpha}^2 + \Psi_{r\beta}^2}, \tag{21}
\]

where

\[
\Psi_r = \frac{|V_r|}{w_r}. \tag{22}
\]

The rotor flux \( (\theta_r) \) angle is calculated by:

\[
\theta_r = \arctan \left( \frac{\Psi_{r\beta}}{\Psi_{r\alpha}} \right). \tag{23}
\]

The estimation of the reactive power and the active power requires prior knowledge of the components of the current and those of the stator voltage. It is retained in the case of a nonlinear command applied to the ASG in particular and especially the command by DRAPC strategy.

Figure 4 shows the logic output deviation \( HC_p \) of the controller following the evolution of the active power \( (P_s) \) compared to the reference active power \( (P_{s_{ref}}) \).

\[
\begin{cases}
+1 & -\Delta P_s/2 \\
0 & 0 \\
+\Delta P_s/2 & -1
\end{cases}
\]

Fig. 4. Active power hysteresis comparator
This corrector makes it possible to command the generator in both directions of rotation, either for a positive or negative active power.

The corrector output, presented by the Boolean variable $HC_p$, indicates directly whether the active power amplitude must be increased in absolute value ($HC_p = 1$) for a positive setpoint and ($HC_p = -1$) for a negative setpoint, or reduced ($HC_p = 0$) this corrector allows a rapid decrease in active power.

This comparator is modeled by the following algorithm, such that $HC_p$ represents the output state of the comparator and $\Delta P_p$ the limit of the hysteresis band

$$
\begin{align*}
\text{if } E_{Ps} < \Delta P_s \text{ so } HC_p &= 1; \\
\text{if } 0 \leq E_{Ps} < \Delta P_s \text{ and } \frac{dE_{Ps}}{dt} > 0 \text{ so } HC_p &= 0; \\
\text{if } 0 \leq E_{Ps} < \Delta P_s \text{ and } \frac{dE_{Ps}}{dt} < 0 \text{ so } HC_p &= 1; \\
\text{if } E_{Ps} < -\Delta P_s \text{ so } HC_p &= -1; \\
\text{if } -\Delta P_s \leq E_{Ps} \leq 0 \text{ and } \frac{dE_{Ps}}{dt} > 0 \text{ so } HC_p &= 0; \\
\text{if } -\Delta P_s \leq E_{Ps} \leq 0 \text{ and } \frac{dE_{Ps}}{dt} < 0 \text{ so } HC_p &= -1,
\end{align*}
$$

(24)

where $E_{Ps} = P_{Sref} - P_s$.

Its purpose to keep the end of the reactive power ($Q_s$) in a circular crown as shown in Fig. 5. The output of the corrector must indicate the direction of evolution of the modulus of $Q_s$, in order to select the corresponding voltage vector.

![Fig. 5. Reactive power hysteresis comparator](image)

For this, a simple two-level hysteresis corrector is ideal, and also allows very good dynamic performance to be obtained.

The output of the corrector is represented by a Boolean variable ($HC_q$) and indicates directly whether the amplitude of the flow must be increased ($HC_q = 1$) or decreased ($HC_q = 0$) in order to maintain:

$$
Q_s^* - Q_s \leq \varepsilon_{Qs} = \Delta Q_s,
$$

(25)

where $Q_s^* = Q_{Sref}$ is the reactive power reference; $\Delta Q_s$ is the corrector hysteresis width; $\varepsilon_{Qs}$ is half the width of the corrector hysteresis band.

The choice of the voltage vector to apply depends on the sign of the error between the reference reactive power $Q_{Sref}$ and the estimated reactive power $Q_s$.

$$
HC_q = 1 \text{ for } E_{Qs} > \varepsilon_{Qs};
HC_q = 0 \text{ for } E_{Qs} < \varepsilon_{Qs}.
$$

(26)

We can write like this:

\[
\begin{align*}
\text{if } E_{Qs} < \Delta Q_s \text{ so } HC_q &= 0; \\
\text{if } 0 \leq E_{Qs} < \Delta Q_s \text{ and } \frac{dE_{Qs}}{dt} > 0 \text{ so } HC_q &= 0; \\
\text{if } 0 \leq E_{Qs} < \Delta Q_s \text{ and } \frac{dE_{Qs}}{dt} < 0 \text{ so } HC_q &= 1; \\
\text{if } E_{Qs} < -\Delta Q_s \text{ so } HC_q &= 0.
\end{align*}
\]

Indeed, if we introduce the difference $\Delta Q_s$ between the reference reactive power ($Q_{Sref}$) and the estimated reactive power ($Q_s$) in a two-level hysteresis comparator (see Fig. 5), it generates at its output the value $HC_q = +1$ to increase the reactive power and $HC_q = 0$ to reduce it, this also allows obtaining a very good dynamic performance of the reactive power [34].

The choice of the voltage vector to apply depends on the sign of the error between the reference reactive power ($Q_{Sref}$) and the estimated flux reactive power.

The Control Panel is built according to the state of the variables $HC_q$ and $HC_p$ and of the zone $Ni$ of the position of $\Phi_s$. It therefore, takes the following form [35] from Table 1.

<table>
<thead>
<tr>
<th>$N_i$</th>
<th>$HC_q$</th>
<th>$HC_p$</th>
<th>1</th>
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<th>3</th>
<th>4</th>
<th>5</th>
<th>6</th>
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<td>0</td>
<td>0</td>
<td>7</td>
</tr>
<tr>
<td>0</td>
<td>1.0</td>
<td>6.1</td>
<td>2</td>
<td>3</td>
<td>4</td>
<td>5</td>
<td>6</td>
<td>1</td>
</tr>
<tr>
<td>1</td>
<td>0.0</td>
<td>7.0</td>
<td>0</td>
<td>0</td>
<td>7</td>
<td>0</td>
<td>0</td>
<td>7</td>
</tr>
<tr>
<td>-1</td>
<td>2.0</td>
<td>3.4</td>
<td>5</td>
<td>6</td>
<td>1</td>
<td>1</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

Table 1

Traditional lookup table of DRAPC technique

where: $HC_q = 0$ reduce the reactive power;
$HC_q = 1$ increase the reactive power;
$HC_p = 1$ increase the active power;
$HC_p = 0$ reduce the active power;
$HC_p = -1$ maintain the active power.

**DRAPC with synergetic STA control.** Direct reactive and active power control has been known to produce a fast response and strong strategy over the electrical generators used in the production of electric current using wind energy. However, there are undulations in active power, torque, reactive power, and current. There are several theories and modern strategies that have been used to minimize the fluctuations that occur in both active and reactive powers. Among these methods, we find: backstepping control, neural networks, neuro-fuzzy control, synergetic control, sliding mode control, and fuzzy logic.

In this paper, a nonlinear DRAPC of ASG is presented based on an SYSTA controller. The variation of rotor and stator resistance due to changes DRAPC controller by introducing errors in the estimated flux linkage, reactive and active powers. The use of the SYSTA strategy minimizes the risks of a risk in the resistance value of the ASG, which leads to a decrease in the undulations.

The DRAPC with SYSTA controllers (SYSTA-DRAPC) is a modification of the classical DRAPC strategy, where the switching table and hysteresis
controllers, have been replaced by a space vector pulse width modulation (SVPWM) technique and SYSTA controllers as shown in Fig. 6. Both of them do not need advanced mathematical models. The DRAPC with SYSTA controller’s goal is to control the active and reactive powers of the ASG. The active power is regulated by the quadrature axis voltage $V_{qdr}^*$, while the reactive power is regulated by the direct axis voltage $V_{drr}^*$.

The sliding surfaces $S(x)$ representing the error between the measured and reference active and reactive powers are given by this relation:

$$S_q = Q_{ref} - Q_s;$$

$$S_p = P_{ref} - P_s.$$

Reactive and active power SYSTA controllers are used to influence respectively on the two rotor voltage components as in (30) and (31)

$$V_{qdr}^* = \alpha_1 \cdot \sqrt{S_p(t)} \cdot \text{sign}(S_p(t)) + \alpha_2 \int \text{sign}(S_p(t)) \cdot \frac{dS_p(t)}{dt} + S_p(t),$$

$$V_{drr}^* = \alpha_1 \cdot \sqrt{S_q(t)} \cdot \text{sign}(S_q(t)) + \alpha_2 \int \text{sign}(S_q(t)) \cdot \frac{dS_q(t)}{dt} + S_q(t).$$

This proposed controller is implemented for a DRAPC technique based on the SYSTA controllers to obtain a minimum active power undulations and to minimize the chattering phenomenon. The controller structure for the SYSTA controllers for the reactive power and active power of the DRAPC technique are presented in Fig. 7 and Fig. 8, respectively.

**Numerical simulations.** The behavior of the structure of the proposed strategies, applied to a high power ASG (1.5 MW), is simulated under the Matlab / Simulink environment and the sampling time is $10^{-5}$ s. The simulation is performed under the following conditions:

The hysteresis band of the reactive power comparator is, in this case, fixed at ±0.05 VAR, and that of the active power comparator at ±0.001 W.

![Fig. 6. DRAPC strategy with SYSTA controllers](image)

![Fig. 7. SYSTA-reactive power controller](image)

![Fig. 8. SYSTA-active power controller](image)

The ASG used in our study has the following parameters: two poles, 50 Hz, 380/696 V, $P_{in} = 1.5$ MW, $L_m = 0.0135$ H, $R_s = 0.012$ $\Omega$, $J = 1000$ kg-m$^2$, $R_t = 0.021$ $\Omega$, $L_t = 0.0137$ H, $L_r = 0.0136$ H, and $f_r = 0.0024$ N m/s [36].

**A. First test.** This first test is the reference tracking test and the results obtained are shown in Fig. 9-18. Figure 9 shows the torque of the proposed and classical strategies. It can be seen, that the amplitudes of the torque depend on the value of the load active power.

Figures 11, 12 represent the active and reactive powers of both strategies. The reactive and active powers track almost perfectly their reference values. Figure 10 shows the current of both DRAPC techniques. It, therefore, confirms that the amplitudes of the currents depend on the value of the load active power and the state of the drive system.

![Fig. 9. Torque $T_e$](image)

![Fig. 10. Stator current $I_{as}$](image)

The zoom in the current, torque, reactive power, and active power is shown in Fig. 13–16, respectively. It can be seen that the proposed strategy minimized the
undulations in current, torque, reactive power, and active power compared to the classical strategy.

Figures 17, 18 show the THD value of the current of both DRAPC strategies. It can be seen through these Figures that the THD value is reduced for the SYSTA-DRAPC (0.19 %) when compared to the classical DRAPC method (1.08 %). On the other hand, this designed strategy minimized the THD value of current compared to other strategies (see Table 2). Based on the results above, it can be said that the SYSTA-DRAPC strategy has proven its efficiency in minimizing undulations and chattering phenomena in addition to keeping the same advantages of the classical DRAPC strategy.

### Table 2

<table>
<thead>
<tr>
<th>Method Name</th>
<th>THD (%)</th>
</tr>
</thead>
<tbody>
<tr>
<td>Ref. [37] Field Oriented Control FOC</td>
<td>3.7</td>
</tr>
<tr>
<td>Ref. [38] Direct Power Control DPC</td>
<td>4.88</td>
</tr>
<tr>
<td>Ref. [39] Virtual-Flux Direct Power Control VFDPC</td>
<td>4.19</td>
</tr>
<tr>
<td>Ref. [39] Sliding Mode Control SMC</td>
<td>3.05</td>
</tr>
<tr>
<td>Ref. [40] Second Order Continuous Sliding Mode</td>
<td>0.98</td>
</tr>
<tr>
<td>– Direct Torque Control SOCSM-DTC</td>
<td></td>
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<tr>
<td>Proposed strategy</td>
<td>1.08</td>
</tr>
<tr>
<td>SYSTA-DRAPC</td>
<td>0.19</td>
</tr>
</tbody>
</table>

### B. Second test.

In this test, we changed the values of both \( L_s \), \( L_r \), \( R_s \), \( R_r \), and \( M \), in order to find out which method is not affected by a change of parameters. The results obtained are shown in Fig. 19–24. Note that there is a change in reactive power, torque, active power, and current due to the fact that both torque and current are related to the changing values of parameters. On the other hand, the classical method was greatly affected by the change of parameters compared to the designed technique (Fig. 25–28), and this is evident in the value of THD (Fig. 19–20). Thus it can be concluded that the DRAPC with proposed SYSTA controllers is more robust than the traditional DRAPC technique.
Conclusions.

In this work, a novel nonlinear command theory for an asynchronous generator was designed based on a synergetic-super twisting algorithm. This command algorithm technique was employed to command the reactive and active powers of the asynchronous generator. The results indicated that the characteristics of the system had improved by using the synergetic-super twisting controllers and that this method could be applied to all electrical generators without exception. This is due to the results obtained. This designed method is easy to apply and does not require any effort, unlike other methods, where it requires mathematical calculations such as sliding mode control and backstepping command, and it is likely has a future in generating electrical energy using electric generators.

Conflict of interest. The authors declare that they have no conflicts of interest.

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